19-3510; Rev 0; 1/05 EVALUATION KIT AVAILABLE 1 700MHz to 2500MHz Variable-Gain Amplifier with Analog Gain Control

General Description

The MAX2057 general-purpose, high-performance variable-gain amplifier (VGA) is designed to operate in the 1700MHz to 2500MHz frequency range*. This device features 15.5dB of gain, 6dB of noise figure, and an output 1dB compression point of 23.8dBm. The MAX2057 also provides an exceptionally high OIP3 level of 37dBm, which is maintained over the entire attenuation range. In addition, the on-chip analog attenuators yield infinite control and high attenuation accuracy over selectable 21dB or 42dB control ranges. Each of these features makes the MAX2057 an ideal VGA for DCS/PCS, cdma2000[™], W-CDMA, and PHS/PAS transmitter and power amplifier AGC circuits.

The MAX2057 is pin compatible with the MAX2056 800MHz to 1000MHz VGA, making this family of amplifiers ideal for applications where a common PC board layout is used for both frequency bands.

The MAX2057 operates from a single +5V supply and is available in a compact 36-pin thin QFN package (6mm x 6mm x 0.8mm) with an exposed paddle. Electrical performance is guaranteed over the extended -40°C to +85°C temperature range.

Applications

DCS 1800/PCS 1900 2G and 2.5G EDGE Base-Station Transmitters and Power Amplifiers

cdmaOne™, cdma2000, Base-Station Transmitters and Power Amplifiers

UMTS/W-CDMA and Other 3G Base-Station Transmitters and Power Amplifiers

PHS/PAS Base-Station Transmitters and Power Amplifiers

Transmitter Gain Control

Receiver Gain Control

Broadband Systems

Automatic Test Equipment

Digital and Spread-Spectrum Communication Systems

Microwave Terrestrial Links

cdmaOne is a trademark of CDMA Development Group. cdma2000 is a registered trademark of Telecommunications Industry Association.

_ Features

- 1700MHz to 2500MHz RF Frequency Range*
- 37dBm Constant OIP3 (Over All Gain Settings)
- 23.8dBm Output 1dB Compression Point
- ♦ 15.5dB Typical Gain at Maximum Gain Setting
- 0.5dB Gain Flatness Over 100MHz Bandwidth
- 6dB Noise Figure at Maximum Gain Setting (Using 1 Attenuator)
- Two Gain-Control Ranges: 21dB and 42dB
- Analog Gain Control
- Single +5V Supply Voltage
- Pin Compatible with MAX2056, 800MHz to 1000MHz RF VGA
- External Current-Setting Resistors Provide Option for Operating VGA in Reduced-Power/Reduced-Performance Mode
- Lead-Free Package Available

***Note:** Operation beyond this range is possible, but has not been characterized.

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE	PKG CODE
MAX2057ETX	-40°C to +85°C	36 Thin QFN-EP** 6mm x 6mm	T3666-2
MAX2057ETX-T	-40°C to +85°C	36 Thin QFN-EP** 6mm x 6mm	T3666-2
MAX2057ETX+D	-40°C to +85°C	36 Thin QFN-EP** 6mm x 6mm	T3666-2
MAX2057ETX+TD	-40°C to +85°C	36 Thin QFN-EP** 6mm x 6mm	T3666-2

**EP = Exposed paddle.

+ = Lead (Pb) free.

D = Dry pack.

-T = Tape-and-reel package.

Pin Configuration/Functional Diagram appear at end of data sheet.

Maxim Integrated Products 1

For pricing, delivery, and ordering information, please contact Maxim/Dallas Direct! at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

MAX2057

ABSOLUTE MAXIMUM RATINGS

V _{CC} to GND	0.3V to +5.5V
V _{CNTL} to GND (with V _{CC} applied)	0V to 4.75V
Current into V _{CNTL} pin (V _{CC} grounded)	40mA
All Other Pins to GND	0.3V to (Vcc + 0.3V)
RF Input Power (IN, IN_A, ATTN_OUT,	OUT_A)+20dBm
RF Input Power (AMP_IN)	+12dBm

θ _{JA} (natural convection)	35°C/W
θJA (1m/s airflow)	31°C/W
θ _{JA} (2.5m/s airflow)	29°C/W
θ _{JC} (junction to exposed paddle)	10°C/W
Operating Temperature Range	40°C to +85°C
Storage Temperature Range	65°C to +150°C
Junction Temperature	+150°C
Lead Temperature (soldering, 10s)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

DC ELECTRICAL CHARACTERISTICS

 $(V_{CC} = +4.75V \text{ to } +5.25V, \text{ no RF signals applied, all input and output ports terminated with 50\Omega, T_A = -40°C \text{ to } +85°C, unless otherwise noted. Typical values are at V_{CC} = +5.0V, T_A = +25°C, unless otherwise noted.)$

PARAMETER	CONDITIONS	MIN	ТҮР	MAX	UNITS
Supply Voltage		4.75	5	5.25	V
Supply Current	$R1 = 1.2k\Omega$, $R2 = 2k\Omega$ (Note 1)		180	230	mA
R _{SET1} Current	$R1 = 1.2k\Omega$ (Note 1)		1		mA
R _{SET2} Current	$R1 = 2k\Omega$ (Note 1)		0.6		mA
Gain-Control Voltage Range	(Note 2)	1.0		4.5	V
Gain-Control Pin Input Resistance	$V_{CNTL} = 1V \text{ to } 4.5V$	250	500		kΩ

AC ELECTRICAL CHARACTERISTICS

(Typical Operating Circuit with one attenuator connected, $V_{CC} = +4.75V$ to +5.25V, $T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Typical values are at $V_{CC} = +5.0V$, $R1 = 1.2k\Omega$, $R2 = 2k\Omega$, $P_{OUT} = +5dBm$, $f_{IN} = 2100MHz$, $V_{CNTL} = 1V$, 50Ω system impedance, second attenuator is not connected, $T_A = +25^{\circ}C$, unless otherwise noted.) (Note 3)

PARAMETER	CONDITIONS		MIN	ТҮР	MAX	UNITS	
Frequency Range			1700		2500	MHz	
Gain	$T_A = +25^{\circ}C$		13.5	15.5	17.5	dB	
		$V_{CNTL} = 1V$		+0.9			
	T	$V_{CNTL} = 1.8V$		+0.41			
	$T_A = +25^{\circ}C \text{ to } -40^{\circ}C$	$V_{CNTL} = 2.6V$		+0.09		- dB	
Maximum Gain Variation		$V_{CNTL} = 3.5V$		-0.16			
Maximum Gain Vanation	$T_A = +25^{\circ}C \text{ to } +85^{\circ}C$	$V_{CNTL} = 1V$		-1			
		$V_{CNTL} = 1.8V$		-0.56			
		$V_{CNTL} = 2.6V$		-0.32			
		$V_{CNTL} = 3.5V$		+0.1			
Reverse Isolation				37		dB	
Noise Figure	(Note 4)			6		dB	
Output 1dB Compression Point				+23.8		dBm	
Output 2nd-Order Intercept Point	From maximum gain to 15dB attenuation, measured at f_1 + f_2 (Note 5)			+64		dBm	
Output 3rd-Order Intercept Point	From maximum gain to 15dB attenuation (Note 5)			+37		dBm	

AC ELECTRICAL CHARACTERISTICS (continued)

(Typical Operating Circuit with one attenuator connected, $V_{CC} = +4.75V$ to +5.25V, $T_A = -40^{\circ}C$ to $+85^{\circ}C$, unless otherwise noted. Typical values are at $V_{CC} = +5.0V$, $R1 = 1.2k\Omega$, $R2 = 2k\Omega$, $P_{OUT} = +5dBm$, $f_{IN} = 2100MHz$, $V_{CNTL} = 1V$, 50Ω system impedance, second attenuator is not connected, $T_A = +25^{\circ}C$, unless otherwise noted.) (Note 3)

PARAMETER	CONDITIONS		MIN	ТҮР	МАХ	UNITS
Output 3rd-Order Intercept Point	$T_A = +25^{\circ}C \text{ to } +85^{\circ}C$			-0.83		- dB
Variation Over Temperature	$T_A = +25^{\circ}C$ to $-40^{\circ}C$			-0.6		uВ
2nd Harmonic	From maximum gain to 15dB	attenuation, $P_{OUT} = +5dBm$		-65		dBc
3rd Harmonic	From maximum gain to 15dB	attenuation, $P_{OUT} = +5dBm$		-83		dBc
RE Cain Control Rongo	$f_{RF} = 1.7 GHz$ to 2.2GHz,	One attenuator	17	20.7		dB
RF Gain-Control Range	$V_{CNTL} = 1V \text{ to } 4.5V$	Two attenuators	34	42.4		
RF Gain-Control Slope	$V_{CNTL} = 1.8V$ to 3.5V			-10		dB/V
Maximum RF Gain-Control Slope	Maximum slope vs. gain-cont	rol voltage		-15.2		dB/V
Gain Flatness Over 100MHz Bandwidth	Peak-to-peak for all settings			0.5		dB
Attenuator Switching Time	15dB attenuation change (Note 6)			500		ns
Attenuator Insertion Loss	Second attenuator (IN_A, OUT_A)			2.2		dB
Input Return Loss	Entire band, all gain settings			18		dB
Output Return Loss	Entire band, all gain settings			15		dB
Group Delay	Input/output 50 Ω lines de-embedded			300		ps
Group Delay Flatness Over 100MHz Bandwidth	Peak to peak			20		ps
Group Delay Change vs. Gain Control	$V_{CNTL} = 1V \text{ to } 4V$			-70		ps
Insertion Phase Change vs. Gain Control	$V_{CNTL} = 1V \text{ to } 4V$			50		degrees

Note 1: Total supply current reduces as R_1 and R_2 are increased.

Note 2: Operating outside this range for extended periods may affect device reliability. Limit pin input current to 40mA when V_{CC} is not present.

Note 3: All limits include external component losses, unless otherwise noted.

Note 4: Noise figure increases by approximately 1dB for every 1dB of gain reduction.

Note 5: $f_1 = 2100MHz$, $f_2 = 2101MHz$, +5dBm/tone at OUT.

Note 6: Switching time is measured from 50% of the control signal to when the RF output settles to ±1dB.

MAX2057

setting, $P_{OUT} = +5$ dBm, linearity measured at $P_{OUT} = +5$ dBm/tone, $T_A = +25$ °C, unless otherwise noted.) SUPPLY CURRENT **INPUT RETURN LOSS OUTPUT RETURN LOSS** vs. SUPPLY VOLTAGE vs. RF FREQUENCY vs. RF FREQUENCY 200 0 0 $T_A = +85°C$ 5 5 190 OUTPUT RETURN LOSS (dB) 10 10 INPUT RETURN LOSS (dB) T_A = -40°C SUPPLY CURRENT (mA) $T_A = -40^{\circ}C$ T_A = +85°C 15 15 180 20 20 $T_A = +25^{\circ}C$ $T_A = +25^{\circ}C$ $T_A = +85^{\circ}C$ 170 25 25 30 30 TA +25°C 160 $T_A = -40^{\circ}C$ 35 35 150 40 40 4.750 4.875 5.000 5.125 5.250 1500 1700 1900 2100 2300 2500 1700 1900 2100 2300 2500 1500 SUPPLY VOLTAGE (V) RF FREQUENCY (MHz) RF FREQUENCY (MHz) **REVERSE ISOLATION GAIN vs. GAIN-CONTROL VOLTAGE GAIN vs. RF FREQUENCY** vs. RF FREQUENCY 40 20 19 $T_A = -40^{\circ}C$ $A = -40^{\circ}C$ $T_A = +85^{\circ}C$ 15 17 REVERSE ISOLATION (dB) Г_А = +25°С 10 35 . T_A = +25°C 15 GAIN (dB) GAIN (dB) TA = +25°C 5 $T_A = +85^{\circ}C$ $T_A = +85^{\circ}C$ $T_A = -40^{\circ}C$ 13 0 30 11 -5 -10 9 25 1500 1700 1900 2100 2300 2500 1.0 1.5 2.0 2.5 3.0 3.5 4.0 1500 1700 1900 2100 2300 2500 RF FREQUENCY (MHz) V_{CNTL} (V) RF FREQUENCY (MHz) **INPUT RETURN LOSS OUTPUT RETURN LOSS** vs. RF FREQUENCY vs. RF FREQUENCY **GAIN vs. RF FREQUENCY** 0 0 20 MAXIMUM GAIN 5 5 15 9dB, 12dB, 15dB, 18dB OUTPUT RETURN LOSS (dB) 10 10 (qB) GAIN REDUCTION 10 INPUT RETURN LOSS 15 15 MAX GAIN (gB) 20 20 5 GAIN MAX GAIN, 3dB, 6dB, 9dB, 12dB, 25 15dB, AND 18dB GAIN REDUCTION 25 0 30 30 6dB GAIN REDUCTION -5 18dB GAIN REDUCTION 3dB GAIN REDUCTION 35 35 40 40 -10 1500 1700 1900 2100 2300 2500 1500 1700 1900 2100 2300 2500 1700 1900 2100 2500 1500 2300 RF FREQUENCY (MHz) RF FREQUENCY (MHz) RF FREQUENCY (MHz) ///XI//

(Typical Application Circuit with **one attenuator** connected, $V_{CC} = +5.0V$, $R1 = 1.2k\Omega$, $R2 = 2k\Omega$, $f_{IN} = 2100MHz$, maximum gain

Typical Operating Characteristics

One Attenuator Configuration

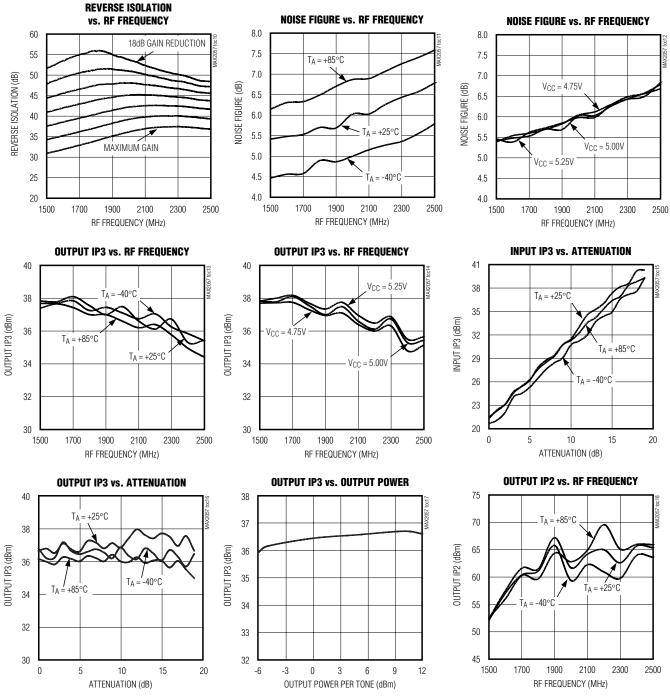
4

Typical Operating Characteristics (continued)

One Attenuator Configuration

MAX2057

(Typical Application Circuit with **one attenuator** connected, $V_{CC} = +5.0V$, R1 = $1.2k\Omega$, R2 = $2k\Omega$, $f_{IN} = 2100MHz$, maximum gain setting, $P_{OUT} = +5dBm$, linearity measured at $P_{OUT} = +5dBm$ /tone, $T_A = +25^{\circ}C$, unless otherwise noted.)



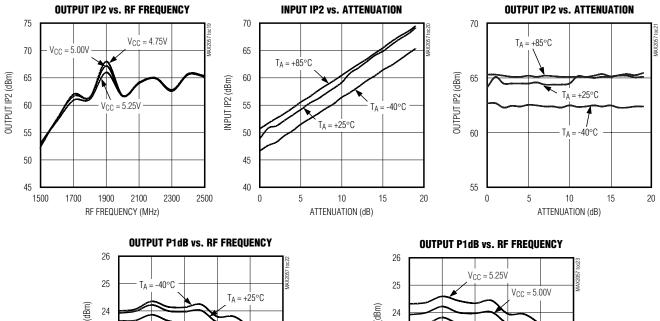
MIXI/M

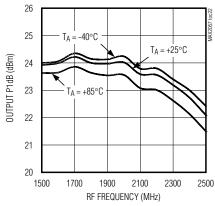
5

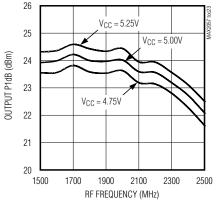
_Typical Operating Characteristics (continued)

One Attenuator Configuration

(Typical Application Circuit with **one attenuator** connected, $V_{CC} = +5.0V$, R1 = $1.2k\Omega$, R2 = $2k\Omega$, $f_{IN} = 2100MHz$, maximum gain setting, $P_{OUT} = +5dBm$, linearity measured at $P_{OUT} = +5dBm$ /tone, $T_A = +25^{\circ}C$, unless otherwise noted.)



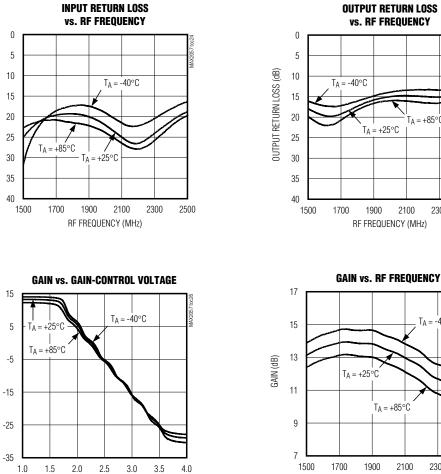




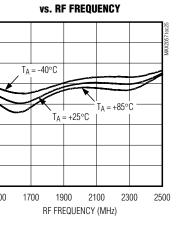
Typical Operating Characteristics

Two Attenuator Configuration

(Typical Application Circuit with two attenuators connected, V_{CC} = +5.0V, R1 = 1.2kΩ, R2 = 2kΩ, f_{IN} = 2100MHz, maximum gain setting, $P_{OUT} = +5$ dBm, linearity measured at $P_{OUT} = +5$ dBm/tone, $T_A = +25^{\circ}$ C, unless otherwise noted.)



V_{CNTL} (V)



T_A = 40°C

2300

2500

. T_A = +85°C

2100

1900

RF FREQUENCY (MHz)

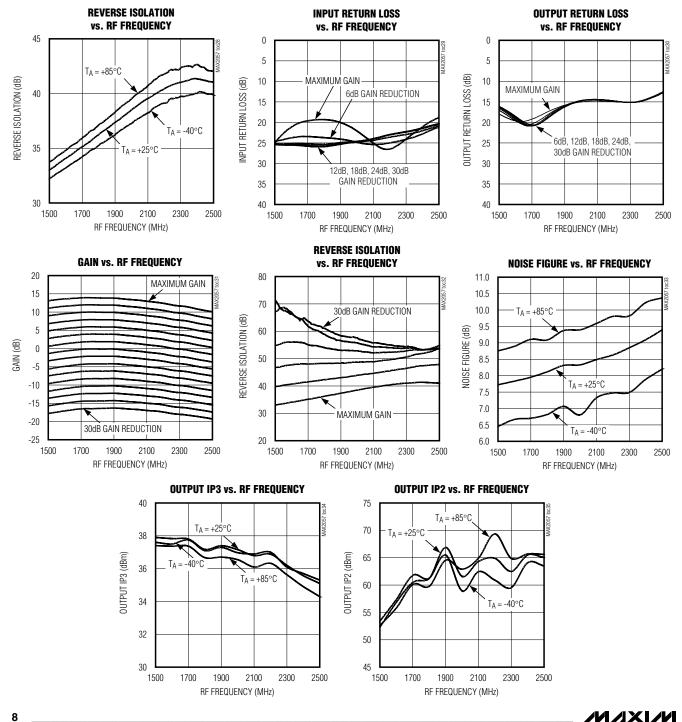
INPUT RETURN LOSS (dB)

GAIN (dB)

Typical Operating Characteristics (continued)

Two Attenuator Configuration

(Typical Application Circuit with two attenuators connected, $V_{CC} = +5.0V$, $R1 = 1.2k\Omega$, $R2 = 2k\Omega$, $f_{IN} = 2100MHz$, maximum gain setting, $P_{OUT} = +5$ dBm, linearity measured at $P_{OUT} = +5$ dBm/tone, $T_A = +25^{\circ}$ C, unless otherwise noted.)



8

_Pin Description

PIN	NAME	FUNCTION	
1, 3, 4, 6, 7, 9, 10, 12, 14, 18, 19, 21–24, 27, 28, 30, 31, 33, 34, 36	GND	Ground. Connect to the board's ground plane using low-inductance layout techniques.	
2	OUT_A	Second-Attenuator Output. Internally matched to 50Ω over the operating frequency band. Connect to IN through a DC-blocking capacitor if greater than 21dB of gain-control range is required. No connection is required if the second attenuator is not used.	
5, 13, 16, 25, 32	V _{CC}	Power Supply. Bypass each pin to GND with capacitors as shown in the <i>Typical Application Circuit</i> . Place capacitors as close to the pin as possible.	
8	IN_A	Second-Attenuator Input. Internally matched to 50Ω over the operating frequency band. Connect to a 50Ω RF source through a DC-blocking capacitor if greater than 21dB of gain-control range is required. No connection is required if the second attenuator is not used.	
11	VCNTL	Analog Gain-Control Input. Limit voltages applied to this pin to a 1V to 4.5V range when V _{CC} is present to ensure device reliability.	
15	R _{SET1}	First-Stage Amplifier Bias-Current Setting. Connect to GND through a 1.2k Ω resistor.	
17	R _{SET2}	Second-Stage Amplifier Bias-Current Setting. Connect to GND through a 2k Ω resistor.	
20	OUT	RF Output. Internally matched to 50Ω over the operating frequency band. Requires a DC-blocking capacitor and a shunt-matching capacitor.	
26	AMP_IN	Amplifier Input. Internally matched to 50Ω over the operating frequency band. Connect to ATTN_OUT through a DC-blocking capacitor.	
29	ATTN_OUT	Attenuator Output. Internally matched to 50Ω over the operating frequency band. Connect to AMP_IN through a DC-blocking capacitor.	
35	IN	RF Input. Internally matched to 50Ω over the operating frequency band. Connect to a 50Ω RF sour through a DC-blocking capacitor if the second attenuator is not used.	
Exposed Paddle	GND	Exposed Paddle Ground Plane. This paddle affects RF performance and provides heat dissipation. This paddle MUST be soldered evenly to the board's ground plane for proper operation.	

MAX2057

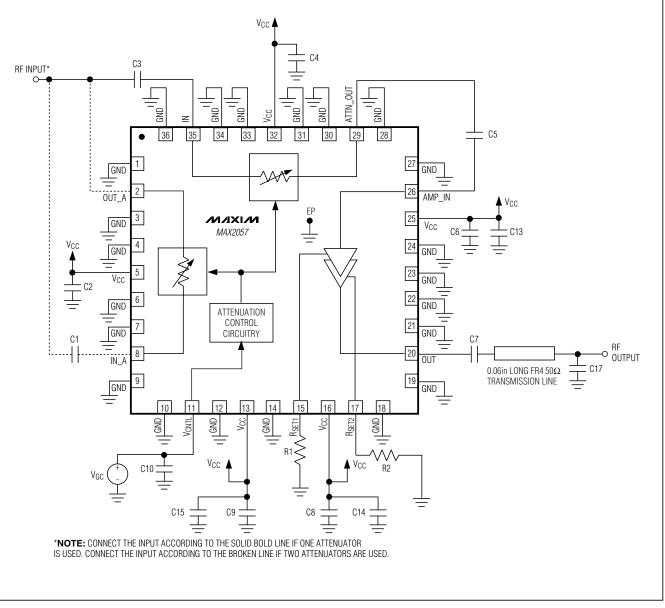


Figure 1. Typical Application Circuit

Detailed Description

The MAX2057 general-purpose, high-performance VGA with analog gain control is designed to interface with 50 Ω systems operating in the 1700MHz to 2500MHz frequency range.

The MAX2057 integrates two attenuators to provide 21dB or 42dB of precision analog gain control, as well

as a two-stage amplifier that has been optimized to provide high gain, high IP3, low noise figure, and low power consumption. The bias current of each amplifier stage can be adjusted by individual external resistors to further reduce power consumption for applications that do not require high linearity.

Table 1. Typical Application CircuitComponent Values

DESIGNATION	VALUE	ТҮРЕ
C1, C3, C5, C7, C10	22pF	Microwave capacitors (0402)
C2, C4, C6, C8, C9	1000pF	Microwave capacitors (0402)
C13, C14, C15	0.1µF	Microwave capacitors (0603)
C17	0.75pF	Microwave capacitor (0402)
R1	1.2kΩ	±1% resistor (0402)
R2	2kΩ	±1% resistor (0402)

Applications Information

Analog Attenuation Control

A single input voltage at the V_{CNTL} pin adjusts the gain of the MAX2057. Up to 21dB of gain-control range is provided through a single attenuator. At the maximum gain setting, each attenuator's insertion loss is approximately 2.2dB. With the single attenuator at the maximum gain setting, the device provides a nominal 15.5dB of cascaded gain and 6dB of cascaded noise figure.

If a larger gain-control range is desired, a second onchip attenuator can be connected in the signal path to provide an additional 21dB of gain-control range. With the second attenuator connected at the maximum gain setting, the device typically exhibits 13.3dB of cascaded gain. Note that the V_{CNTL} pin simultaneously adjusts both on-chip attenuators.

The V_{CNTL} input voltage drives a high-impedance load (>250k Ω). It is suggested that a current-limiting resistor be included in series with this connection to limit the input current to less than 40mA should the control voltage be applied when V_{CC} is not present. A series resistor of greater than 200 Ω will provide complete protection for 5V control voltage ranges. Limit V_{CNTL} input voltages to a 1.0V to 4.5V range when V_{CC} is present to ensure the reliability of the device.

Amplifier Bias Current

The MAX2057 integrates a two-stage amplifier to simultaneously provide high gain and high IP3. Optimal performance is obtained when R1 and R2 are equal to $1.2k\Omega$ and $2k\Omega$, respectively. The typical supply current is 180mA and the typical output IP3 is 37dBm under these conditions.

Increasing R₁ and R₂ from the nominal values of 1.2k Ω and 2k Ω reduces the bias current of each amplifier stage, which reduces the total power consumption and IP3 of the device. This feature can be utilized to further decrease power consumption for applications that do not require high IP3.

Layout Considerations

A properly designed PC board is an essential part of any RF/microwave circuit. Keep RF signal lines as short as possible to reduce losses, radiation, and inductance. For best performance, route the ground-pin traces directly to the exposed pad underneath the package. This pad **MUST** be connected to the ground plane of the board by using multiple vias under the device to provide the best RF and thermal conduction path. Solder the exposed pad on the bottom of the device package to a PC board exposed pad.

Power-Supply Bypassing

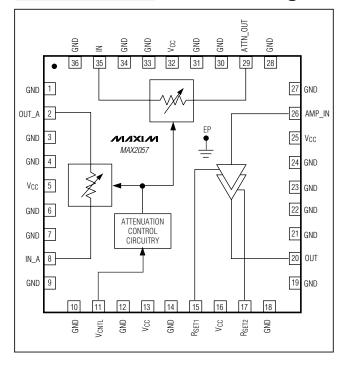
Proper voltage-supply bypassing is essential for highfrequency circuit stability. Bypass each VCC pin with capacitors placed as close to the device as possible. Place the smallest capacitor closest to the device. Refer to the MAX2057 evaluation kit data sheet for more details.

Exposed Paddle RF and Thermal Considerations

The EP of the MAX2057's 36-pin thin QFN-EP package provides a low-thermal-resistance path to the die. It is important that the PC board on which the IC is mounted be designed to conduct heat from this contact. In addition, the EP provides a low-inductance RF ground path for the device.

The EP **MUST** be soldered to a ground plane on the PC board either directly or through an array of plated via holes. Soldering the pad to ground is also critical for efficient heat transfer. Use a solid ground plane wherever possible.

MAX2057



Pin Configuration/ Functional Diagram Chip Information

TRANSISTOR COUNT: 5191 PROCESS: BICMOS

_Package Information

For the latest package outline information, go to **www.maxim-ic.com/packages**.

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.



_____Maxim Integrated Products, 120 San Gabriel Drive, Sunnyvale, CA 94086 408-737-7600

© 2005 Maxim Integrated Products Printed USA MAXIM is a registered trademark of Maxim Integrated Products, Inc.